# Correspondence 

# Nonintersecting Subspaces Based on Finite Alphabets 

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#### Abstract

Two subspaces of a vector space are here called "nonintersecting" if they meet only in the zero vector. Motivated by the design of noncoherent multiple-antenna communications systems, we consider the following question. How many pairwise nonintersecting $\boldsymbol{M}_{\boldsymbol{t}}$-dimensional subspaces of an $\boldsymbol{m}$-dimensional vector space $V$ over a field $\mathbb{F}$ can be found, if the generator matrices for the subspaces may contain only symbols from a given finite alphabet $\mathcal{A} \subseteq \mathbb{F}$ ? The most important case is when $\mathbb{F}$ is the field of complex numbers $\mathbb{C}$; then $M_{t}$ is the number of antennas. If $\mathcal{A}=\mathbb{F}=\mathbf{G F}(q)$ it is shown that the number of nonintersecting subspaces is at most $\left(q^{m}-1\right) /\left(q^{M_{t}}-1\right)$, and that this bound can be attained if and only if $\boldsymbol{m}$ is divisible by $\boldsymbol{M}_{\boldsymbol{t}}$. Furthermore, these subspaces remain nonintersecting when "lifted" to the complex field. It follows that the finite field case is essentially completely solved. In the case when $\mathbb{F}=\mathbb{C}$ only the case $M_{t}=2$ is considered. It is shown that if $\mathcal{A}$ is a PSK-configuration, consisting of the $2^{r}$ complex roots of unity, the number of nonintersecting planes is at least $2^{r(m-2)}$ and at most $2^{r(m-1)-1}$ (the lower bound may in fact be the best that can be achieved).


Index Terms-Multiple-antenna communications, noncoherent systems, nonintersecting subspaces, space-time codes.

## I. Introduction

The problem studied in this correspondence is motivated by communication over multiple antenna channels [10], [21], [22]. In [6] and [22], it was shown that the capacity of the multiple-antenna channel grows linearly as a function of the minimum of the numbers of transmitting and receiving antennas. The proof assumed that the receiver has complete information about the channel. In [21], the emphasis was placed on reducing error probability by introducing correlation between signals transmitted from different antennas. These points of view can be combined by observing that there is a tradeoff between rate and reliability [21], [23].

Most of the early work on multiple-antenna communications assumed that the receiver was able to track the channel perfectly-i.e., used coherent detection. If coherent detection is difficult or too expensive, one can use noncoherent detection, as studied in [10]. The main result from this work is that the capacity is still (almost) linear in the minimum number of transmitting or receiving antennas [10], [26]. Hence, both in the coherent and noncoherent cases, it was established

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that the use of multiple-antennas leads to a gain in information transmission rate.

In [11], the error probability of multiple-antenna noncoherent communication channels was investigated. It was shown there (and in [26]) that if the channel is not known to the receiver, the coding problem is equivalent to one of packing subspaces (which represent codewords) according to a certain notion of distance. If an orthonormal basis for the subspaces were used, the diversity order [the slope of the error probability with respect to signal-to-noise ratio was shown to depend on the dimension of the intersection of the subspaces.

In particular, to obtain maximal diversity, one wishes to construct a family of subspaces which intersect only at the origin. By a slight abuse of notation we will say that two vector spaces are "nonintersecting" if their only common point is the zero vector. An extensive characterization and classification of such group differential space-time code constructions was given in [19]. The focus of much of this work is on constructing codes which have the nonintersecting subspace property without imposing any constraints on the number of different symbols used to define the codewords-that is, the codewords are allowed to use a signal constellation that is larger than the minimum possible.

Motivated by these observations, the main question addressed in the present correspondence is the construction of nonintersecting subspaces, subject to the constraint that the codewords are defined using symbols from a fixed, small constellation. We focus on two cases: one in which the symbols are taken from a finite field and the other where they are taken from a phase-shift keying (PSK) arrangement, i.e., are complex roots of unity. Our aim is to find constructions that give the largest number of nonintersecting subspaces (i.e., have the highest rate) subject to these constraints. Note that the constructions given in this correspondence could result in nonunitary space-time codes.

It is worth remarking that a recent paper by Lusina et al. [16] discusses an analogous problem for the case of coherent decoders. Another related paper by Lu and Kumar [15] explores code constructions with fixed alphabet constraints for achieving different points on the rate-diversity tradeoff. Again, only coherent decoders are considered. A very recent paper by Kammoun and Belfiore [13] directly addresses the problem of constructing codes for noncoherent systems with a large value of $\Lambda\left(\boldsymbol{X}, \boldsymbol{X}^{\prime}\right)$ [see (7)] between subspaces. However, their approach is quite different from ours.

The present correspondence is organized as follows. In Section II, we establish notation and formalize the question being studied. In Section III, we study the case when the symbols are taken from a finite field, and in Section IV when they are complex roots of unity (i.e., PSK constellations). Section V compares the different constructions and mentions some directions for further research.

## II. PreLiminaries

In order to motivate the question studied in this correspondence and establish notation, consider a multiple antenna channel. Let the number of transmitting antennas be $M_{t}$ and the number of receiving antennas be $M_{r}$. If $\boldsymbol{y}(k) \in \mathbb{C}^{M_{r}}$ is the received (column) vector at time $k$, we can write

$$
\begin{equation*}
\boldsymbol{y}(k)=\sqrt{E_{s}} \boldsymbol{H}(k) \boldsymbol{x}(k)+\boldsymbol{z}(k) \tag{1}
\end{equation*}
$$

where the matrix $\boldsymbol{H}(k) \in \mathbb{C}^{M_{r} \times M_{t}}$ represents the channel, the column vector $\boldsymbol{x}(k) \in \mathbb{C}^{M_{t}}$ is the channel input, $E_{s}$ is the signal power per
transmitting antenna, and $\boldsymbol{z}(k) \in \mathbb{C}^{M_{r}}$ is zero mean i.i.d. Gaussian noise with $\mathbb{E}\left[\boldsymbol{z}(k) \boldsymbol{z}(k)^{H}\right]=N_{0} \boldsymbol{I}$. We assume a Rayleigh flat fading model, i.e., that the elements of $\boldsymbol{H}(k)$ are i.i.d. with a zero mean complex Gaussian distribution of unit variance. The channel is assumed to be block time-invariant, that is, $\boldsymbol{H}(k)$ is independent of $k$ over a transmission block of $m$ symbols, say $\boldsymbol{H}(k)=\boldsymbol{H}$ (although $\boldsymbol{H}(k)$ may vary from block to block). Looking at a single block of length $m$, during which the channel is assumed to be time-invariant, we can write

$$
\begin{align*}
\boldsymbol{Y} & =[\boldsymbol{y}(1), \ldots, \boldsymbol{y}(m)] \\
& =\sqrt{E_{s}} \boldsymbol{H}[\boldsymbol{x}(1), \ldots, \boldsymbol{x}(m)]+[\boldsymbol{z}(1), \ldots, \boldsymbol{z}(m)] \\
& =\sqrt{E_{s}} \boldsymbol{H} \boldsymbol{X}+\boldsymbol{Z} . \tag{2}
\end{align*}
$$

The focus of this correspondence is on constructing the space-time codewords $\boldsymbol{X}$, subject to the constraint that the elements of $\boldsymbol{X}$ are selected from a particular alphabet $\mathcal{A}$.

## A. Criteria for Code Design

In this correspondence, we assume that the receiver will not attempt to estimate the channel matrix $\boldsymbol{H}$, i.e., that we have a noncoherent receiver. Therefore, the maximum likelihood detection rule without using the channel state information [11] is that we should decode $Y$ as that codeword $\hat{\boldsymbol{X}}$ which maximizes

$$
\begin{equation*}
\frac{\exp \left(-\operatorname{Trace}\left[\boldsymbol{Y} \Psi^{-1} \boldsymbol{Y}^{H}\right]\right)}{|\pi \Psi|^{M_{r}}} \tag{3}
\end{equation*}
$$

where $\Psi=\boldsymbol{I}+E_{s} \boldsymbol{X}^{H} \boldsymbol{X}, H$ denotes the transposed complex conjugate or adjoint matrix, and $|\cdot|$ denotes a determinant. Using the matrix inversion lemma ([12,, p. 19]) and the property that $|\boldsymbol{I}+\boldsymbol{A B}|=$ $|\boldsymbol{I}+\boldsymbol{B A}|$, the detection criterion can be rewritten as

$$
\begin{align*}
& \hat{\boldsymbol{X}}=\arg \max _{X}\left\{E_{s} \operatorname{Trace}\left[\boldsymbol{Y} \boldsymbol{X}^{H}\left(\boldsymbol{I}+E_{s} \boldsymbol{X} \boldsymbol{X}^{H}\right)^{-1} \boldsymbol{X} \boldsymbol{Y}^{H}\right]\right. \\
&\left.\quad-\log \left|\boldsymbol{I}+E_{s} \boldsymbol{X} \boldsymbol{X}^{H}\right|\right\} \tag{4}
\end{align*}
$$

In the absence of channel state information at the receiver, Hochwald and Marzetta [11] argue that, at high signal-to-noise ratio, one should use unitary codewords $\boldsymbol{X}$, satisfying $\boldsymbol{X} \boldsymbol{X}^{H}=m \boldsymbol{I}$. Using this in (4), it follows that $\hat{\boldsymbol{X}}$ should be chosen to maximize

$$
\begin{equation*}
\text { Trace }\left[\boldsymbol{Y} \boldsymbol{X}^{H} \boldsymbol{X} \boldsymbol{Y}^{H}\right] \tag{5}
\end{equation*}
$$

This implies that, for unitary codewords, the decoder should project the received signal onto the subspace defined by each of the codewords and declare the codeword with the maximal projection to be the winner. In [11] it is shown that, for unitary codewords, the probability that a transmitted codeword $\boldsymbol{X}$ is decoded as the codeword $\hat{\boldsymbol{X}}$ is bounded above by

$$
\begin{equation*}
\frac{1}{\left|\boldsymbol{I}_{M_{t}}+\frac{\rho^{2} m^{2}}{4(1+\rho m)}\left[\boldsymbol{I}_{M_{t}}-\frac{1}{m^{2}} \hat{\boldsymbol{X}} \boldsymbol{X}^{H} \boldsymbol{X} \hat{\boldsymbol{X}}^{H}\right]\right|^{M_{r}}} \tag{6}
\end{equation*}
$$

where $\rho=\frac{E_{s}}{N_{0}}$ is the signal-to-noise ratio. If the signal-to-noise ratio is large, this pairwise error probability behaves like $\left(\frac{\Lambda \rho}{4}\right)^{-M_{r} \nu}$, where $\nu$ is the rank of $\left[\boldsymbol{I}_{M_{t}}-\frac{1}{m^{2}} \hat{\boldsymbol{X}} \boldsymbol{X}^{H} \boldsymbol{X} \hat{\boldsymbol{X}}\right]$

$$
\begin{aligned}
\Lambda & =\Lambda(\boldsymbol{X}, \hat{\boldsymbol{X}}) \\
& =\left|m \boldsymbol{I}_{M_{t}}-\frac{1}{m} \hat{\boldsymbol{X}} \boldsymbol{X}^{H} \boldsymbol{X} \hat{\boldsymbol{X}}^{H}\right|_{+}^{\frac{1}{\nu}}
\end{aligned}
$$

and $|\cdot|_{+}$denotes the product of the nonzero eigenvalues. Note that

$$
\left|\left[\begin{array}{l}
\boldsymbol{X} \\
\hat{\boldsymbol{X}}
\end{array}\right]\left[\begin{array}{ll}
\boldsymbol{X}^{H} & \hat{\boldsymbol{X}}^{H}
\end{array}\right]\right|=\left|m^{2} \boldsymbol{I}_{M_{t}}-\hat{\boldsymbol{X}} \boldsymbol{X}^{H} \boldsymbol{X} \hat{\boldsymbol{X}}^{H}\right|
$$

which shows that $\nu=M_{t}$ is equivalent to the condition that the rows of $\boldsymbol{X}, \hat{\boldsymbol{X}}$ are linearly independent [11]. For this to happen we must have $m \geq 2 M_{t}$.

Another interpretation can be given in terms of the principal angles between subspaces corresponding to pairs of codewords. The principal angles between subspaces $\boldsymbol{X}$ and $\boldsymbol{X}^{\prime}$ are given by $\cos \theta_{i}=\frac{1}{m} \sigma_{i}\left(\boldsymbol{X}^{\prime} \boldsymbol{X}^{H}\right)$ where $\sigma_{i}(\cdot)$ is the $i$-th singular value of the matrix ([4], [7]). Using this we obtain

$$
\begin{equation*}
\Lambda\left(\boldsymbol{X}, \boldsymbol{X}^{\prime}\right)=m \prod_{i=1}^{\nu}\left[1-\cos ^{2} \theta_{i}\right]=m \prod_{i=1}^{\nu} \sin ^{2} \theta_{i} \tag{7}
\end{equation*}
$$

This provides a better measure of how good a code is: not only should the subspaces be nonintersecting, the value of $\Lambda\left(\boldsymbol{X}, \boldsymbol{X}^{\prime}\right)$ should be large for every pair $\boldsymbol{X}, \boldsymbol{X}^{\prime}$ of distinct subspaces. The error probability will be dominated by the pair of codewords with the least rank $\nu$ and the least "distance" $\Lambda\left(\boldsymbol{X}, \boldsymbol{X}^{\prime}\right)$. For well separated subspaces this "distance" can also be approximated by

$$
\begin{equation*}
\sum_{i=1}^{\nu} \sin ^{2} \theta_{i} \tag{8}
\end{equation*}
$$

which is the the notion of distance between subspaces used in [4] and [2].

Another way to compare these codes is by using the notion of diversity order (cf. [21]).

Definition II.1: If the average error probability $\bar{P}_{e}(\rho)$ as a function of the signal-to-noise ratio $\rho$ satisfies

$$
\begin{equation*}
\lim _{\rho \rightarrow \infty} \frac{\log \left(\bar{P}_{e}(\rho)\right)}{\log (\rho)}=-d \tag{9}
\end{equation*}
$$

the coding scheme is said to have diversity order $d$.
It follows from (6) that the diversity order of the unitary space-time coding scheme is equal to $M_{r} \nu$. The maximal diversity order that can be achieved is, therefore, $M_{r} M_{t}$. We call codes that achieve this bound fully diverse codes.

In brief, to get a diversity order of $M_{r} M_{t}$, we need to construct nonintersecting subspaces which are far apart in the metric defined by (7). Though this property of nonintersecting subspaces yielding fully diverse codes was demonstrated for unitary codewords [11], nonunitary constructions with this property also yield fully diverse codes using the appropriate detection rule given in (4). This can be seen from the following argument. Consider a single receive antenna ( $M_{r}=1$ ). In the absence of noise, the received signal $\boldsymbol{Y}=\sqrt{E_{s}} \boldsymbol{H} \boldsymbol{X}$, and since $\boldsymbol{H} \in \mathbb{C}^{1 \times M_{t}}$ the received vector $\boldsymbol{Y} \in$ rowspan $(\boldsymbol{X})$. Therefore, unless $H$ is zero, we can distinguish the different codewords as long as the subspaces are nonintersecting. In applying this argument to the high signal-to-noise ratio regime, we just need $\|\boldsymbol{H}\|^{2} \gg 1 /$ SNR for us to be able to distinguish nonintersecting subspaces in the presence of noise. Since for $H \sim \mathbb{C} \mathcal{N}\left(0, \boldsymbol{I}_{M_{t}}\right)$ and high signal-to-noise ratio, we have

$$
\mathbb{P}\left[\|\boldsymbol{H}\|^{2}<1 / \mathrm{SNR}\right] \approx 1 / \mathrm{SNR}^{M_{t}}
$$

[23], we see that nonintersecting subspaces would yield fully diverse codes. This argument can be made precise, but that is beyond the scope of this correspondence since we only use noncoherent multiple antenna systems to motivate the problem of nonintersecting subspaces with a finite alphabet.

Given this motivation, we will focus on obtaining maximal diversity order by constructing families of subspaces which are nonintersecting. Note that some of the constructions could yield nonunitary space-time codes. In order to further improve performance we need to maximize $\Lambda\left(\boldsymbol{X}, \boldsymbol{X}^{\prime}\right)$ over all pairs $\boldsymbol{X}, \boldsymbol{X}^{\prime}$ of distinct subspaces. The rate of a code $C$ is $R=\frac{1}{m} \log (|\mathcal{C}|)$. In trying to construct the maximal number of
nonintersecting subspaces, we attempt to get the highest rate codes that achieve maximal diversity order.

## B. Statement of the Problem

Definition II.2: Let $\mathbb{F}$ be a field. A codeword or subspace will mean an $M_{t}$-dimensional subspace of $\mathbb{F}^{m}$. Two subspaces $\Pi_{1}$ and $\Pi_{2}$ are said to be nonintersecting over $\mathbb{F}$ if their intersection is trivial, i.e., if $\Pi_{1} \cap \Pi_{2}=\{0\}$.

Suppose $\Pi_{1}$ is generated by (row) vectors $u_{1}, \ldots, u_{M_{t}} \in \mathbb{F}^{m}$, and $\Pi_{2}$ is generated by vectors $v_{1}, \ldots, v_{M_{t}} \in \mathbb{F}^{m}$. Let $P:=\left[\begin{array}{l}\Pi_{1} \\ \Pi_{2}\end{array}\right]$ denote the $2 M_{t} \times m$ matrix with rows $u_{1}, \ldots, u_{M_{t}}, v_{1}, \ldots, v_{M_{t}}$. Then the following lemma is readily established.

Lemma II.1: The following properties are equivalent: i) $\Pi_{1}$ and $\Pi_{2}$ are nonintersecting; ii) $P$ has rank $2 M_{t}$ over $\mathbb{F}$; and iii) if $m=2 M_{t}$ the determinant of $P$ is nonzero.

Suppose now that instead of allowing the entries in the matrices $\Pi_{1}$ and $\Pi_{2}$ to be arbitrary elements of $\mathbb{F}$, we restrict them to belong to a finite subset $\mathcal{A} \subseteq \mathbb{F}$, called the alphabet. In other words, the vectors $u_{1}, \ldots, u_{M_{t}}, v_{1}, \ldots, v_{M_{t}}$ must belong to $\mathcal{A}^{m}$. The question that we address is the following: given $M_{t}, m$ and a finite alphabet $\mathcal{A} \subseteq \mathbb{F}$, how many subspaces can we find which are generated by vectors from $\mathcal{A}^{m}$ and which are pairwise nonintersecting over $\mathbb{F}$ ? Furthermore, if the size of $\mathcal{A}$ is specified in advance, which choice of $\mathcal{A}$ permits the biggest codes?

We first dispose of the trivial case when $M_{t}=1$. Two nonzero vectors $u, v$ are said to be projectively distinct over a field $\mathbb{F}$ if there is no $a \in F$ such that $u=a v$. Then if $M_{t}=1$, the maximum number of nonintersecting subspaces is simply the maximum number of projectively distinct vectors in $\mathcal{A}^{m}$.

In the following sections we will investigate the first question for two kinds of alphabets: a) $\mathcal{A}$ is a finite field $\mathbb{F}$ (Section III), and b) $M_{t}=2$ and $\mathcal{A} \subseteq C^{m}$ is a set of complex roots of unity (Section IV).

Of course, for the application to multiple-antenna code design, the subspaces need to be disjoint over $\mathbb{C}$. In Section III, Theorem III.4, we translate the results obtained over $\mathbb{F}$ to this case by "lifting" the subspaces to the complex field. Furthermore, for this application, the case $m=2 M_{t}$ is the most important.

## III. Finite Fields

In this section we assume that the alphabet $\mathcal{A}$ and the field $\mathbb{F}$ are both equal to the finite field $\operatorname{GF}(q)$, where $q$ is a power of a prime $p$. At the end of the section we show how to "lift" these planes to the complex field (see Theorem III.4). In this case, there is an obvious upper bound which can be achieved in infinitely many cases. Let $V$ denote the vector space $\mathrm{GF}(q)^{m}$.

Theorem III.1: The number of pairwise nonintersecting $M_{t}$-dimensional subspaces of $V$ is at most

$$
\begin{equation*}
\frac{q^{m}-1}{q^{M_{t}}-1} \tag{10}
\end{equation*}
$$

Proof: There are $q^{m}-1$ nonzero vectors in $V$ and each subspace contains $q^{M_{t}}-1$ of them. No nonzero vector can appear in more than one subspace.

It is convenient here to use the language of projective geometry, cf. [17, Appendix B]. Recall that the points of the projective space $P(s, q)$ are equivalence classes of nonzero vectors from $\operatorname{GF}(q)^{s+1}$, where two vectors are regarded as equivalent if one is a nonzero scalar multiple of the other.

A spread [9] in $P G(s, q)$ is a partition of the points into copies of $P G(r, q)$.

Theorem III.2: Such a spread exists if and only if $r+1$ divides $s+1$.
Proof: This is a classical result, due to André ([1]; [9, Theorem 4.1.1]).

Corollary III.3: The bound (10) can be attained whenever $M_{t}$ divides $m$, and only in those cases.

Proof: This is immediate from the theorem, since a set of points in a projective space represents a set of projectively distinct lines in the corresponding vector space.

Note that the condition is independent of $q$. If a set of nonintersecting subspaces meeting (10) exists over one finite field then it exists over every finite field.

Furthermore, it is straightforward to construct the nonintersecting subspaces meeting the bound in (10), as we now show. The nonzero elements of a finite field $\mathbb{F}$ form a multiplicative group which will be denoted by $\mathbb{F}^{*}$. This is a cyclic group [14, Ch. 2].

Suppose $M_{t}$ divides $m$, and consider the fields $F_{0}=\mathrm{GF}(q), F_{1}=$ $\mathrm{GF}\left(q^{M_{t}}\right), F_{2}=\mathrm{GF}\left(q^{m}\right)$. Then $F_{0} \subseteq F_{1} \subseteq F_{2}$. $\operatorname{By}$ regarding $\operatorname{GF}\left(q^{m}\right)$ as a vector space of dimension $m$ over $\operatorname{GF}(q)$ we can identify $F_{2}$ with $V$. Similarly we can regard $F_{1}$ as a $M_{t}$-dimensional subspace of $V$. The desired spread is now obtained by partitioning $F_{2}^{*}$ into (multiplicative) cosets of $F_{1}^{*}$.

Example III.1: We consider the case $M_{t}=2, m=4$, and $\mathcal{A}=$ $\mathrm{GF}(2)=\{0,1\}$. Then $F_{0}=\mathrm{GF}(2), F_{1}=\mathrm{GF}(4), F_{2}=\mathrm{GF}(16)$. Each plane in $\operatorname{GF}(2)^{4}$ contains three nonzero vectors, and $\operatorname{GF}(2)^{4}$ itself contains 15 nonzero vectors. We wish to find a spread of $P G(1,2)$ s inside $P G(3,2)$, that is, a partitioning of the 15 vectors into five disjoint sets of three, where each set of three adds to the zero vector.

Let $\operatorname{GF}(16)=\operatorname{GF}(2)[\alpha]$, where $\alpha^{4}+\alpha+1=0$. A table of the elements of this field and their binary representations can be found for example in [17, Fig. 3.3]. Then $\operatorname{GF}(4)$ is the subfield $\left\{1, \alpha^{5}, \alpha^{10}\right\}$, so $F_{1}^{*}=\left\{\alpha^{5}, \alpha^{10}\right\}$, and we obtain the desired partition

$$
F_{2}^{*}=\bigcup_{j=0}^{4} \alpha^{j} F_{1}^{*}
$$

Only two of the three vectors are needed to define each plane, so we have the following generators for the five planes

$$
(1, \alpha),\left(\alpha, \alpha^{6}\right),\left(\alpha^{2}, \alpha^{7}\right),\left(\alpha^{3}, \alpha^{8}\right),\left(\alpha^{4}, \alpha^{9}\right) .
$$

Using the table in [17], we convert these to explicit generator matrices for the five nonintersecting planes

$$
\left[\begin{array}{l}
1000 \\
0110
\end{array}\right],\left[\begin{array}{l}
0100 \\
0011
\end{array}\right],\left[\begin{array}{l}
0010 \\
1101
\end{array}\right],\left[\begin{array}{l}
0001 \\
1010
\end{array}\right],\left[\begin{array}{l}
1100 \\
0101
\end{array}\right]
$$

The problem is therefore essentially solved as long as $M_{t}$ divides $m$. If not, we can use partial spreads-see the surveys in [5] and [20].

We end this section by observing that a set of nonintersecting subspaces over a finite field $\mathcal{A}=\operatorname{GF}(q), q=p^{k}, p$ prime, can always be "lifted" to a set of nonintersecting subspaces over a complex alphabet $\overline{\mathcal{A}}$ of the same size.

This can be done as follows. Suppose $\operatorname{GF}(q)=\operatorname{GF}(p)[\alpha]$, where $\alpha$ is a root of a primitive irreducible polynomial $f(X) \in \mathrm{GF}(p)[X]$. Let $n=p^{k}-1$ and let $\mu_{n}=e^{2 \pi i / n}$. Adjoining $\mu_{n}$ to the rational numbers $\mathbb{Q}$, we obtain the cyclotomic field $\mathbb{Q}\left(\mu_{n}\right)$, with ring of integers $\mathbb{Z}\left[\mu_{n}\right]$. It is a classical result from number theory that the ideal $(p)$ in $\mathbb{Z}\left[\mu_{n}\right]$ factors into $g=\varphi(n) / k$ distinct maximal prime ideals $\mathfrak{p}_{1}, \mathfrak{p}_{2}, \ldots, \mathfrak{p}_{\mathfrak{g}}$, where $\varphi(\cdot)$ is the Euler totient function. Furthermore, for each $\mathfrak{p}_{\mathfrak{j}}$, the residue class ring $\mathbb{Z}\left[\mu_{n}\right] / \mathfrak{p}_{\mathfrak{j}} \cong \mathrm{GF}(\mathfrak{q})$ (see, for example, [3, Theorem 10.45], [18, Ch. 10, Sec. 3B], [24, Theorem 2.13], [25, Theorem 7-2-4]). If we choose $\mathfrak{p}_{\mathfrak{j}}$ to be the ideal generated by $p$ and $f\left(\mu_{n}\right)$, then $\mathbb{Z}\left[\mu_{n}\right] / \mathfrak{p}_{\mathfrak{j}}$ is exactly the version of $\mathrm{GF}(q)$ that we started with. Note that since $\mathfrak{p}_{\mathfrak{j}}$ contains $(p)$, it acts as reduction $\bmod p$ on $\mathbb{Z}$.

We therefore have a ring homomorphism from $\mathbb{Z}\left[\mu_{n}\right]$ to $\mathrm{GF}(q)$ given by

$$
\begin{equation*}
\phi: \mathbb{Z}\left[\mu_{n}\right] \xrightarrow{\bmod p} \rightarrow \mathbb{Z}\left[\mu_{n}\right] / \mathfrak{p}_{\mathfrak{j}} \xrightarrow{\cong} \mathfrak{G F}(\mathfrak{q}) \tag{11}
\end{equation*}
$$

In this way, we can lift vectors over $\operatorname{GF}(q)$ to vectors over the alphabet $\bar{A}$ consisting of 0 and the $q-1$ powers of $\mu_{n}$.

Example: Let $\mathrm{GF}(8)=\mathrm{GF}(2)[\alpha]$ where $\alpha$ is a root of $X^{3}+X+1$. Then $q=8, n=7, \mu_{7}=e^{2 \pi i / 7}$. To lift $\mathrm{GF}(8)$ to $\mathbb{C}$ we write $\mathrm{GF}(8)=$ $\left\{0,1, \alpha, \alpha^{2}, \ldots, \alpha^{6}\right\}$, and lift 0 to 0 and $\alpha^{j}$ to $\mu_{7}^{j}$ for $j=0, \ldots, 6$.
Let $\Pi$ be an $M_{t}$-dimensional subspace of $\mathrm{GF}(q)^{m}$. By lifting each element of a generator matrix we obtain an $M_{t}$-dimensional subspace $\bar{\Pi} \subseteq \mathbb{C}^{m}$, defined over an alphabet $\overline{\mathcal{A}}$ of size $q$.

Theorem III.4: If two subspaces $\Pi_{1}, \Pi_{2}$ of $\mathrm{GF}(q)^{m}$ are nonintersecting, so are their lifts $\bar{\Pi}_{1}, \bar{\Pi}_{2}$.

Proof: Let $P:=\left[\begin{array}{l}\Pi_{1} \\ \Pi_{2}\end{array}\right]$ and $\bar{P}:=\left[\begin{array}{l}\bar{\Pi}_{1} \\ \bar{\Pi}_{2}\end{array}\right]$. By Lemma II.1, $P$ has a $2 M_{t} \times 2 M_{t}$ invertible submatrix. Since $\phi$ is a ring homomorphism, the lift of this submatrix is also invertible.

It follows that the subspaces constructed in Corollary III. 3 are also nonintersecting when lifted to the complex field.

This construction gives full diversity order noncoherent space-time codes when the elements of the codewords are restricted to belong to a finite field. Their rate is

$$
R=\frac{1}{m} \log \left(q^{m}-1\right)-\frac{1}{m} \log \left(q^{M_{t}}-1\right)<\log (q)
$$

which, according to Theorem III.1, is the maximal achievable rate for diversity order $M_{t} M_{r}$. Moreover, this relationship implies that for fully diverse codes constructed from a finite field, we cannot achieve a rate higher than $\log (|\mathcal{A}|)$.

## IV. PSK Constellations

Throughout this section, we assume that the alphabet $\mathcal{A}$ consists of the set of complex $2^{r}$ th roots of unity, that is, $\mathcal{A}=\left\{e^{2 \pi i j / 2^{r}}, 0 \leq\right.$ $\left.j<2^{r}\right\}$, for some $r \geq 1$. Let $\mu=e^{2 \pi i / 2^{r}}$ be a primitive $2^{r}$ th root of unity; $\mathcal{A}$ is a cyclic multiplicative group with generator $\mu$. Here, we assume that $M_{t}=2$, that is, the code consists of a set of pairwise nonintersecting planes.

Example IV.1: Some examples of roots of unity are as follows:

1. If $r=1, \mu=-1$ and the alphabet is $\mathcal{A}=\{1,-1\}$.
2. If $r=2, \mu=i$ and the alphabet is $\mathcal{A}=\{1, i,-1,-i\}$.
3. If $r=3, \mu=(1+i) / \sqrt{2}$ and the alphabet is $\mathcal{A}=\left\{e^{\pi i j / 4}, 0 \leq\right.$ $j \leq 7\}$. This is the 8 -PSK constellation.
There is a trivial upper bound.
Theorem IV.1: Let $\mathcal{A}$ be the set of $2^{r}$ roots of unity, $r \geq 1$. Then the number of pairwise nonintersecting planes is at most $\frac{1}{2}|\mathcal{A}|^{m-1}=$ $2^{(m-1) r-1}$.

Proof: If $v_{1}, v_{2} \in \mathcal{A}^{m}$ are the generators for a plane, that plane also contains all multiples $\mu^{j} v_{1}$ and $\mu^{j} v_{2}$, a total of $2|\mathcal{A}|$ vectors. Since these sets of vectors must all be disjoint, the number of planes is at most $|\mathcal{A}|^{m} /(2|\mathcal{A}|)$.

The same argument shows that there are at most $\frac{1}{M_{t}}|\mathcal{A}|^{m-1}$ nonintersecting $M_{t}$-dimensional subspaces of complex $m$-dimensional space for any finite alphabet $A$. The implication of this in terms of rate is that

$$
R \leq \frac{m-1}{m} \log (|\mathcal{A}|)-\frac{1}{m} \log \left(M_{t}\right)<\log (|\mathcal{A}|)
$$

Hence, for fully diverse codes constructed from PSK constellations, we cannot achieve a rate exceeding $\log (|\mathcal{A}|)$.

Example IV.2: Let $\mathcal{A}$ be the set $\{1, i,-1,-i\}$ and take $m=4$. The total number of vectors in $\mathcal{A}^{4}$ is $4^{4}$. Each vector has four multiples, so each plane accounts for at least eight vectors. Therefore, there are at most $\frac{4^{4}}{8}=32$ planes.

In the other direction we will prove the following.
Theorem IV.2: Assume $r \geq 1$ and that $m \geq 2$ is even. There exist $N=|\mathcal{A}|^{m-2}=2^{(m-2) r}$ pairwise nonintersecting planes in $\mathbb{C}^{m}$ defined using the complex $2^{r}$ th roots of unity.

Note that the upper and lower bounds coincide in the case $r=1$, that is, when $\mathcal{A}=\{1,-1\}$.

The proof is simplified by the use of valuations (cf. [8]). If $x \in$ $\mathbb{Q}, x=2^{a} \frac{b}{c}$ with $a, b, c \in \mathbb{Z}, c \neq 0, b$, and $c$ odd, then the 2-adic valuation of $x$ is $\nu_{2}(x)=a$. Similarly, suppose $x$ belongs to the cyclotomic field $\mathbb{Q}(\mu)$. Since $1-\mu$ is a prime in $\mathbb{Z}[\mu]$, we can write $x$ uniquely as $(1-\mu)^{a} \frac{b}{c}$ with $a \in \mathbb{Z}, b, c \in \mathbb{Z}[\mu], c \neq 0, b$ and $c$ relatively prime to $1-\mu$. The $(1-\mu)$-adic valuation of $x$ is then $\nu_{1-\mu}(x)=a$. It is easy to check that for $k \in \mathbb{Z}, k \neq 0, \nu_{1-\mu}\left(1-\mu^{k}\right)=2^{\nu_{2}(k)}$. In particular, if $k \in \mathbb{Z}$ is odd, $\nu_{1-\mu}\left(1-\mu^{k}\right)=1$.

We will also need a lemma.
Lemma IV.3: Let $\Pi$ be a plane in $\mathbb{C}^{m}$ generated by vectors $v_{1}, v_{2}$, and denote by

$$
\tilde{\Pi}_{1}=\left[\begin{array}{lll}
v_{1} & x_{11} & x_{12} \\
v_{2} & x_{21} & x_{22}
\end{array}\right]
$$

and

$$
\tilde{\Pi}_{2}=\left[\begin{array}{lll}
v_{1} & y_{11} & y_{12} \\
v_{2} & y_{21} & y_{22}
\end{array}\right]
$$

two different embeddings of $\Pi$ into $\mathbb{C}^{m+2}$. Then $\tilde{\Pi}_{1} \cap \tilde{\Pi}_{2}=\{0\}$ if and only if

$$
\left|\begin{array}{ll}
y_{11}-x_{11} & y_{12}-x_{12} \\
y_{21}-x_{21} & y_{22}-x_{22}
\end{array}\right| \neq 0
$$

Proof: By Lemma II.1, it is necessary and sufficient that the matrix $P:=\left[\begin{array}{l}\tilde{\Pi}_{1} \\ \tilde{\Pi}_{2}\end{array}\right]$ have rank 4. Subtracting the first and second rows of $P$ from the third and fourth rows, we get the matrix

$$
\left[\begin{array}{ccc}
v_{1} & x_{11} & x_{12} \\
v_{2} & x_{21} & x_{22} \\
0 & y_{11}-x_{11} & y_{12}-x_{12} \\
0 & y_{21}-x_{21} & y_{22}-x_{22}
\end{array}\right]
$$

and the result follows.
We now give the proof of the theorem, for which we use induction on even values of $m$. For $m=2$ we take the single plane

$$
\left[\begin{array}{cc}
1 & 1 \\
1 & -1
\end{array}\right]
$$

Suppose the result is true for $m$. For each of the $|\mathcal{A}|^{m-2}$ pairwise nonintersecting planes in $\mathbb{C}^{m}$ we will construct $|\mathcal{A}|^{2}$ planes in $\mathbb{C}^{m+2}$, such that full set of planes so obtained is pairwise nonintersecting; this will establish the desired result.

If two planes are nonintersecting in $\mathbb{C}^{m}$ then they are certainly nonintersecting when embedded in any way in $\mathbb{C}^{m+2}$. So we need only show that the $|\mathcal{A}|^{2}$ embeddings of any single plane are pairwise nonintersecting.

Let $\Pi$ be a plane in $\mathbb{C}^{m}$ generated by vectors $v_{1}, v_{2}$, and denote by $\tilde{\Pi}(a, b)$ the plane in $\mathbb{C}^{m+2}$ with generator matrix

$$
\left[\begin{array}{ccc}
v_{1} & \mu^{a} & \mu^{b} \\
v_{2} & \mu^{a+b} & \mu^{a+2 b+1}
\end{array}\right]
$$

for $a, b=0,1, \ldots, 2^{r}-1$.
We will use Lemma IV. 3 to show that all the planes $\{\tilde{\Pi}(a, b) \mid a \in$ $\mathcal{A}, b \in \mathcal{A}\}$ are pairwise nonintersecting. For this we must show that

$$
\left|\begin{array}{cc}
\mu^{c}-\mu^{a} & \mu^{d}-\mu^{b} \\
\mu^{c+d}-\mu^{a+b} & \mu^{c+2 d+1}-\mu^{a+2 b+1}
\end{array}\right|=0
$$

if and only if $a=c$ and $b=d$.
The above determinant is equal to

$$
\begin{align*}
\mu^{2 c+2 d+1}\left(1-\mu^{a-c}\right) & \left(1-\mu^{(a-c)+2(b-d)}\right) \\
& -\mu^{c+2 d}\left(1-\mu^{b-d}\right)\left(1-\mu^{(a-c)+(b-d)}\right) . \tag{12}
\end{align*}
$$

If the determinant is zero, the $(1-\mu)$-adic valuations of the two terms on the right must be equal, that is

$$
\begin{equation*}
2^{\nu_{2}(a-c)}+2^{\nu_{2}(a-c+2(b-d))}=2^{\nu_{2}(b-d)}+2^{\nu_{2}(a-c+b-d)} \tag{13}
\end{equation*}
$$

We must show that this is true if and only if $a=c$ and $b=d$. We consider four cases, depending on the parity of $a-c$ and $b-d$. If $a-c \equiv$ $1, b-d \equiv 1(\bmod 2)$ then $(12)$ reads $1+1=1+2^{\nu_{2}(a-c+b-d)} \geq 3$ (since $a-c+b-d$ is even), a contradiction. Similarly, if $a-c \equiv 1$, $b-d \equiv 0(\bmod 2)$ we get $1+1=2^{\nu_{2}(b-d)}+1$, and if $a-c \equiv$ $0, b-d \equiv 1(\bmod 2)$ we get $2^{\nu_{2}(a-c)}+2^{\nu_{2}(a-c+2(b-d))}=1+1$, which are also contradictions. The fourth possibility is $a-c \equiv b-d \equiv$ $0(\bmod 2)$. Let $a-c=2^{s} x$ and $b-d=2^{t} y$, where $x$ and $y$ are odd, $s, t \geq 1$. We have

$$
\nu_{2}(a-c+2(b-d))= \begin{cases}s & \text { if } s<t \\ s & \text { if } s=t \\ \geq t & \text { if } s>t\end{cases}
$$

and

$$
\nu_{2}(a-c+2(b-d))= \begin{cases}s & \text { if } s<t \\ \geq s & \text { if } s=t \\ t & \text { if } s>t\end{cases}
$$

Substituting these valuations in (13) again gives a contradiction. This concludes the proof of Theorem IV.2.

## V. Discussion

Table I compares the codes constructed in Sections III and IV in the case $M_{t}=2$, i.e., codes which are pairwise nonintersecting 2-D subspaces of $\mathbb{C}^{m}$, for $m=4,6$ and 8 , and alphabets $\mathcal{A}$ of sizes 2,4 , and 8 . The top entry in each cell gives the number of planes obtained from the finite field construction (Corollary III.3). The bottom entry gives the lower and upper bounds obtained using complex $|\mathcal{A}|$ th roots of unity, from Theorem IV. 2 and Theorem IV.1. Asymptotically, the rates of the two constructions are very similar. Both satisfy $\log ($ number of codewords $) / m \approx \log (|\mathcal{A}|)$, for $m$ large, and so both asymptotically achieve the maximal rate possible for fully diverse codes.

Note that the construction via finite fields results in codes for which alphabet consists of 0 and the complex $(|A|-1)$ st roots of unity, whereas the construction via PSK constellations produces codes in which the symbols are the complex $|A|$ th roots of unity (and 0 is not used).

We end by mentioning some topics for further research.

- We also used clique-finding algorithms to search for larger sets of planes than those given in Theorem IV.2, again taking $\mathcal{A}$ to

TABLE I
Number of Pairwise Nonintersecting Planes in $\mathbb{C}^{m}$ for Various Sizes of the Alphabet $|\mathcal{A}|$ (See Text for Details)

|  | $m=4$ | $m=6$ | $m=8$ |
| :--- | :---: | :---: | :---: |
| $\|\mathcal{A}\|=2$ | 5 | 21 | 85 |
|  | $4-4$ | $16-16$ | $64-64$ |
| $\|\mathcal{A}\|=4$ | 17 | 273 | 4369 |
|  | $16-32$ | $256-512$ | $4096-8192$ |
| $\|\mathcal{A}\|=8$ | 65 | 4161 | 266305 |
|  | $64-256$ | $4096-16384$ | $262144-1048576$ |

be the set of $2^{r}$ th complex roots of unity. These searches were unsuccessful, and so we have not mentioned them elsewhere in the correspondence. These negative results lead us to conjecture, albeit weakly, that the lower bounds in Theorem IV. 2 cannot be improved. It would be nice to have a better upper bound than that in Theorem IV. 1 for the case $r>1$. It would also be a worthwhile project to do a more extensive computer search for better codes, both for the above alphabet and for other alphabets.

It is straightforward to formulate the search as a clique-finding problem. The first step is to prepare a list of candidate subspaces, making sure that the generator matrices use only symbols from $\mathcal{A}$, and that the subspaces have the specified dimension and are distinct (a subspace may have many different generator matrices: only one version is placed on the list of candidates). Then a graph is constructed with the candidate subspaces as vertices, and with an edge joining two vertices if and only if the subspaces are nonintersecting. Then a good code is a maximal clique in this graph.

- Can the construction in Theorem IV. 2 be generalized to the case when $M_{t}$ is larger than 2? In particular, it would be interesting to do a computer search in the case $M_{t}=3$ and $m=6$.
- This correspondence has focused only on the existence and construction of finite alphabet codes which achieve maximal diversity order, and we did not consider decoding complexity. The decoding problem involves weighted projections of the received matrix $\boldsymbol{Y}$ onto the candidate subspaces [see (4)]. In general, this may require a search over $2^{m R}$ codewords, where $R$ is the rate of the code. Since this number grows exponentially with the code length, a natural question to ask is whether there are codes which are optimally decodable in polynomial time, or have polynomial time suboptimal decoders which perform satisfactorily.
- In [4] (see also [2]) a large number of optimal or putatively optimal packings of subspaces in $\mathbb{C}^{m}$ were constructed using (8) as a measure of "distance" between subspaces. It would be worthwhile repeating these calculations using (7) instead.


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# The Asymptotic Capacity of Multiple-Antenna Rayleigh-Fading Channels 

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#### Abstract

We consider the asymptotic behavior of the capacity of mul-tiple-antenna Rayleigh-fading channels in the limit as the transmit and receive arrays become large. We show that the capacity converges in distribution to a Gaussian random variable, and give closed-form formulas for its mean and variance. These results enable us to derive the first asymptotic formula for outage rates, as well as a sharper estimate of the error in previously reported asymptotic formulas for ergodic capacity. Although these formulas are asymptotic, we show by simulation that they are often quite accurate, even for relatively small arrays.


Index Terms-Capacity, fading channels, multiple-antenna communication, random matrix theory, space-time coding.

## I. INTRODUCTION

The seminal work of Telatar [8] and Foschini and Gans [3] demonstrated that using multiple antennas at the transmitter and receiver can substantially increase the capacity of fading multipath channels. The input distribution that maximizes the mutual information is known to be independent and identically distributed (i.i.d.) Gaussian. For a particular channel realization $H$, the mutual information in nats per second per Hertz $(\mathrm{nps} / \mathrm{Hz})$ is given by

$$
\begin{equation*}
C(t, r, \rho)=\ln \left|I+\frac{\rho}{t} H H^{\dagger}\right| \tag{1}
\end{equation*}
$$

that depends upon the channel fading path gains $H$, the signal-to-noise ratio (SNR) $\rho$, and the numbers of transmit and receive antennas $t$ and $r$, respectively. $C(t, r, \rho)$ may be viewed as a random variable defined on the ensemble of all possible channel matrices. However, the actual operational limit on reliable communication depends on how $H$ evolves with time: If the path gains are ergodic, $C(t, r, \rho)$ can be averaged over many channel realizations, and the corresponding operational limit is called the ergodic capacity

$$
\begin{equation*}
C_{e}(t, r, \rho)=\mathcal{E}[C(t, r, \rho)] \tag{2}
\end{equation*}
$$

If the path gains are static, however, we observe only one channel realization and there is often an irreducible outage probability associated with every positive rate of transmission. For outage probability $0<q<1$, reliable communication is limited by the outage rate

$$
\begin{equation*}
C_{q}(t, r, \rho)=\sup \{R \geq 0: \operatorname{Pr}[C(t, r, \rho)<R] \leq q\} \tag{3}
\end{equation*}
$$

Ergodic capacity and outage rates are usually estimated by Monte Carlo methods, since closed-form formulas are not known, except for a few special cases. Telatar [8] obtained a general formula for ergodic capacity in terms of an integral of a series of Laguerre polynomials.

[^0]
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